

## GENERAL DESCRIPTION

The PT2312E is a high performance AC/DC power switch for charger and adaptor applications. Due to the PSR control, the PT2312E can provide accurate constant voltage (CV) and constant current (CC) regulation without opto-coupler and TL431.

The PT2312E adopts PFM method in CC control mode and PFM/PWM method in CV control mode. In order to achieve high performance, The PT2312E uses quasi-resonant (QR) switching to improve the efficiency, frequency shuffling technique to optimize EMI design and built-in cable drop compensation to meet good load regulation.

The PT2312E utilizes soft start control and multi-protection to meet high reliability, such as cycle-by-cycle current limiting, output OVP, VCC OVP/UVLO, sense resistor open/short protection, OTP etc.

PT2312E is offered in SOP-7 package.

## FEATURES

- PSR without opto-coupler and TL431
- 5% CV and 5% CC regulation at universal AC input
- QR operation for high efficiency
- Frequency shuffling technique for EMI
- Built-in cable drop compensation
- Built-in adaptive current peak regulation
- Built-in primary side winding inductance compensation
- Built-in leading edge blanking (LEB)
- Power on soft-start
- Cycle-by-cycle current limiting
- VCC OVP/UVLO, Output OVP
- CS resistor open and short protection
- Over temperature protection(OTP)
- Auto-recovery after trigger protection

## APPLICATIONS

- Adapters/Chargers
- LED Lighting
- Auxiliary Power Supplies

## ORDERING INFORMATION

PACKAGE	TEMPERATURE RANGE	ORDERING PART NUMBER	TRANSPORT MEDIA	MARKING
SOP-7	-40°C to 85°C	PT2312EESOG	Tape and Reel 2500 units	 PT2312E xxxxxX

## TYPICAL APPLICATION CIRCUIT

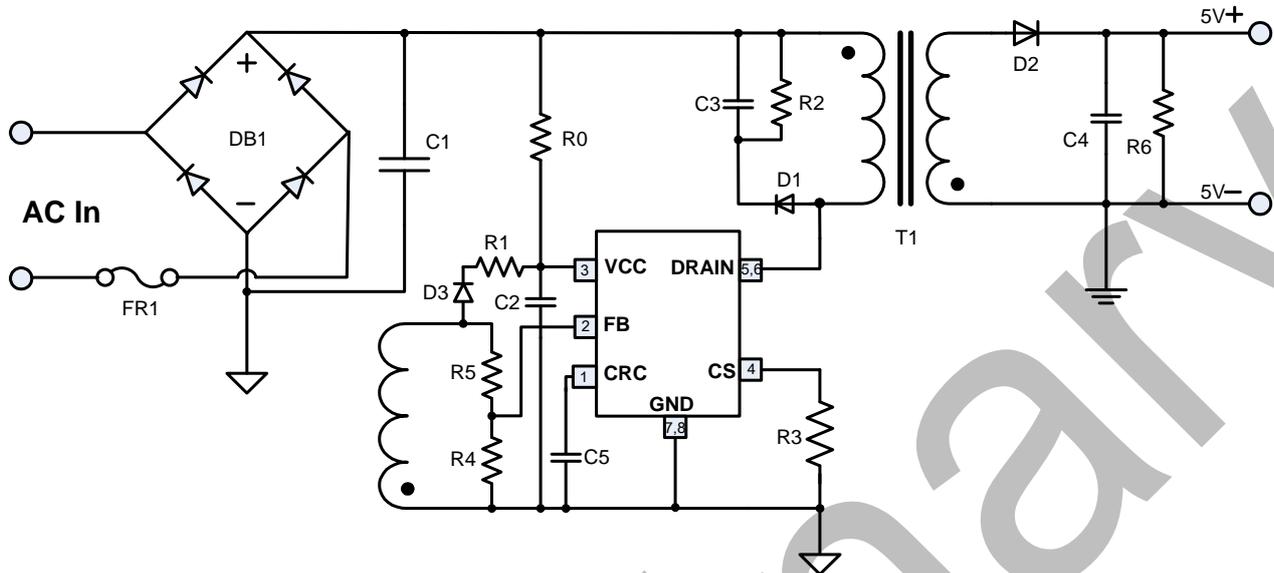


Figure1. Typical Application of PT2312E

## PIN ASSIGNMENT

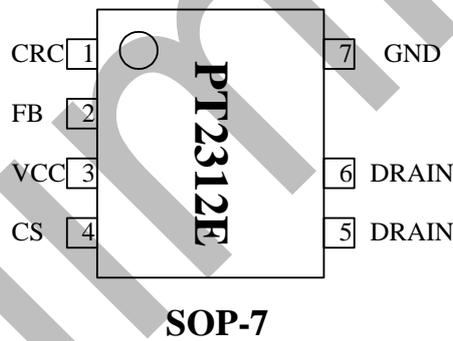


Figure 2. Pin Configuration of PT2312E (Top View)

## PIN DESCRIPTIONS

PIN No.	PIN NAMES	DESCRIPTION
1	CRC	This pin connects a capacitor (10nF is recommended) to GND for system stability regulation
2	FB	Voltage feedback, The CV and CC regulation are realized based on the voltage sampling of this pin
3	VCC	The power supply for the IC. In order to get the correct operation of the IC, a capacitor with low ESR should be placed as close as possible to the VCC pin
4	CS	Primary Side Peak Current Detection Port
5,6	DRAIN	Open-drain of MOSFET
7	GND	The ground of the IC

### ABSOLUTE MAXIMUM RATINGS (Note1)

SYM	PARAMETER	VALUE	UNIT
V <sub>CC</sub>	Voltage at VCC to GND	-0.3 to 28	V
	Voltage at Drain to GND	0 to 650	V
	Voltage at FB, CS, CRC to GND	-0.3 to 6	V
T <sub>J</sub>	Operating Junction Temperature	-40 to 150	°C
T <sub>STG</sub>	Storage Temperature	-55 to 150	°C
T <sub>LEAD</sub>	Lead Temperature (Soldering, 10 sec)	300	°C
θ <sub>JA</sub>	Thermal Resistance (Junction to Ambient)	184	°C/W
	ESD (Human Body Model)	2500	V

**Note 1:** Stresses greater than those listed under “Absolute Maximum Ratings” may cause permanent damages to the device. Functional operation of the device at these or any other conditions beyond those indicated under “Recommended Operating Conditions” is not implied. For maximum safe operating conditions, refer to Electrical Characteristics.

### SIMPLIFIED BLOCK DIAGRAM

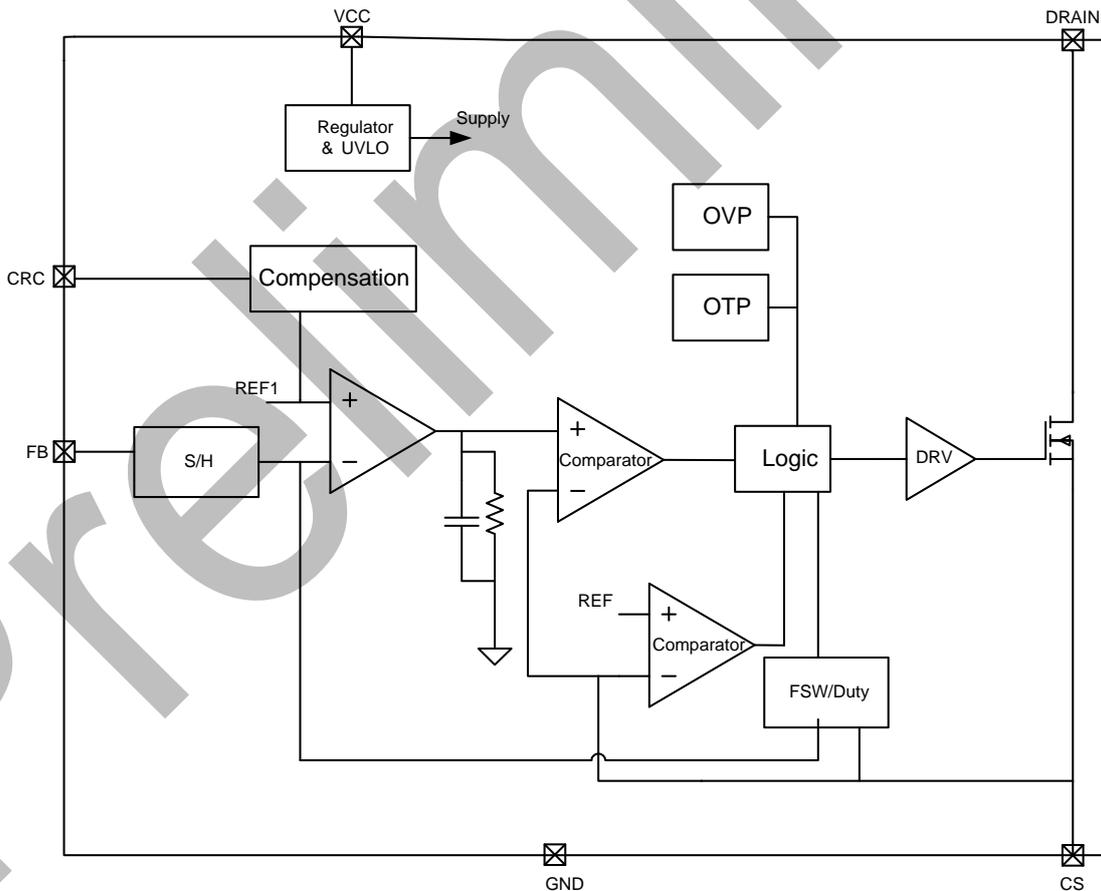


Figure 3. Simplified Block Diagram of PT2312E

## ELECTRICAL CHARACTERISTICS

(TA=25°C, VCC=20.5V, unless otherwise specified)

SYMBOL	PARAMETER	TEST CONDITION	MIN	TYP	MAX	UNIT
<b>Supply Voltage (VCC) Section</b>						
VCC <sub>ON</sub>	Turn-on threshold	VCC rising	18.5	20.5	22.5	V
VCC <sub>OFF</sub>	Turn-off threshold	VCC falling	7.7	8.2	8.7	V
VCC <sub>OVP</sub>	Over voltage protection threshold	Ramp VCC until gate shut down	23	24	25	V
VCC <sub>clamp</sub>	Clamp voltage	I <sub>VCC</sub> =10mA	25.5	27	28	V
ICC <sub>ST</sub>	Start-up current	Before turn on, VCC=18V			1	μA
ICC <sub>OP</sub>	Operation current	After VCC turn on			1	mA
ICC <sub>FAULT</sub>	Protection current	Fault conditions		500		μA
<b>Feedback (FB) Section</b>						
VFB	Feedback Voltage	Operation at CV mode	2.475	2.5	2.525	V
VFB <sub>OVP</sub>	FB OVP			3		V
IFB <sub>LEAK</sub>	FB leakage current				0.1	μA
I <sub>cable</sub>	Cable compensation current		40	42.5	45	μA
<b>Current Sense (CS) Section</b>						
V <sub>CS</sub>	CS current limit		0.485	0.5	0.515	V
T <sub>LEB</sub>	LEB time			300		ns
<b>Frequency Section</b>						
F <sub>MIN</sub>	Minimum frequency	No load		500		Hz
Δf/Freq	Frequency shuffling range			7		%
<b>OTP Section</b>						
T <sub>OTP</sub>	OTP temperature			160		°C
T <sub>recovery</sub>	Recovery temperature			120		°C
<b>Power MOSFET Section</b>						
BV <sub>DSS</sub>	Drain-Source breakdown Voltage		650			V
R <sub>DSON</sub>	On Resistance	Static, Id=2.0A		4.0	4.8	Ω

### OPERATION DESCRIPTION

The PT2312E is a high performance AC/DC power controller for charger and adaptor applications. Due to the PSR control, the PT2312E can provide accurate constant voltage (CV) and constant current (CC) regulation without opto-coupler and TL431, as shown in Figure 4.

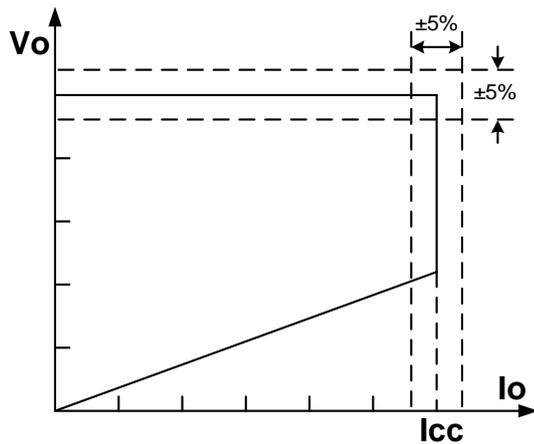


Figure 4. Typical CC/CV Curve

#### ● Start up Current and Start up Control

The VCC of PT2312E could be charged up above UVLO threshold and starts up quickly because of the very low startup current. A large value startup resistor can therefore be used to minimize the power loss in application. In addition, soft-start is designed to assure the reliability.

#### ● Operating Current

The operating current of PT2312E is about 1mA. Good efficiency is achieved with the low operating current together with 'Multi-mode' control features. Also, it is very important for low standby power design.

#### ● Constant Voltage Operation

The PT2312E is designed for flyback system,

which always works in DCM mode at both CC/CV control (Refer to Figure 1).

The controller works in PFM/PWM mode at CV operation. And the output voltage of system can be sensed via the divider of auxiliary winding. During primary MOSFET turn-on time, the secondary output diode is reversed, then the load current is supplied from the output capacitors. The current in the primary winding ramps up. Until MOSFET turns off, the primary current transfers to the secondary side and the output diode become conducting state. The waveform of auxiliary voltage reflecting the output voltage is shown in figure 5 and the equation is given by

$$V_{AUX} = \frac{N_{AUX}}{N_s} \cdot (V_o + V_d) \quad (1)$$

Where  $V_d$  indicates the forward voltage on the output diode,  $N_{AUX}/N_s$  is the turn ratio.

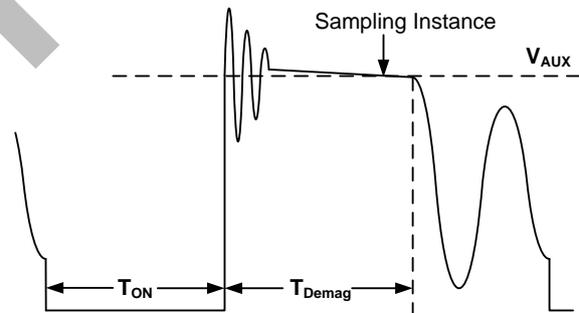


Figure 5. Auxiliary voltage waveform

The FB pin will detect the output voltage through the divider connected on the auxiliary winding, the auxiliary voltage is sampled at the two thirds of the de-magnetization time and will be hold until the next sampling. The FB signal will be compared with a internal reference ( $V_{ref}$ ) to get a error signal, which is used for voltage loop feedback at CV operation mode.

### ● Constant Current Operation

When sampled voltage on FB pin is below Vref, the voltage feedback loop is not work and the system now works in CC operation mode. The switching frequency of system depends on the output voltage. This moment, the CS threshold (Vcs) will be kept at a constant value because of the saturation of the voltage loop, thus the constant output current can be achieved. The CC point and maximum output power can be adjusted by the external current sense resistor Rcs at CS pin, which is shown in typical application diagram. The output power is adjusted through CC point change. If a larger Rcs is selected, a smaller CC point will be get, meanwhile the smaller output power will be became to, and vice versa. Figure 6 shows the secondary current waveform on transformer.

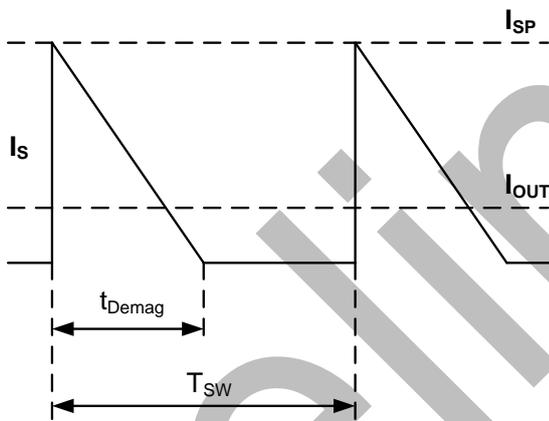


Figure 6. Secondary Current Waveform

According to the above, the output constant current is given by:

$$I_p = \frac{V_{CS}}{R_{CS}} \quad (2)$$

$$I_{out} = \frac{1}{2} \times I_{SP} \times \frac{t_{Demag}}{T} \quad (3)$$

$$I_{SP} = \frac{N_p}{N_s} \times I_p \quad (4)$$

So :

$$I_{out} = 0.25 \times \frac{N_p}{N_s} \times \frac{V_{CS}}{R_{CS}} \quad (5)$$

Where:  $t_{Demag}/T$  is set to be 0.5 internally.

### ● Quasi-Resonant Operation

The PT2312E incorporates a unique proprietary quasi-resonant switching scheme that achieves valley-mode turn on for every PWM/PFM switching cycle in CV operation. This unique feature greatly reduces the switching loss and dv/dt across the entire operating range of the power supply. Due to the nature of quasi-resonant switching, the actual switching frequency can vary slightly cycle by cycle, providing the additional benefit of reducing EMI. Together, this innovative control architecture makes the PT2312E to achieve highest overall efficiency and lowest EMI, without causing audible noise over entire operating range.

### ● Operation switching frequency

The switching frequency is not set by external components, but depends on the load condition and the operation mode, it is adaptively controlled. For flyback operating in DCM, The maximum output power is given by

$$P_{OMAX} = \frac{1}{2} L_p F_{sw} I_p^2 \quad (6)$$

Where  $L_p$  indicates the inductance of primary winding and  $I_p$  is the peak current of primary winding.

According to the equation 6, when the switching frequency (Fsw) is set, the maximum output power is proportional to the inductance of the primary winding as well as the square of the primary peak current. In order to compensate the change from variations of primary winding inductance, the switching frequency is locked by an internal loop, so that the switching frequency is

$$F_{SW} = \frac{1}{2t_{Demag}} \quad (7)$$

Here,  $t_{Demag}$  is the de-magnetization time, which is shown in figure 5. In CC mode, according to the equation 5, since the  $V_{CS}$  is an internal constant, the output current is proportional to the turn ratio and inversely proportional to the value of  $R_{cs}$ , it will not change with primary winding inductance changes. For a fixed system, the switching frequency is proportional to the output voltage when working in CC mode.

### ● Frequency shuffling for EMI

The frequency shuffling (switching frequency modulation) is implemented in PT2312E. The oscillation frequency is modulated so that the tone energy is spread out. The spread spectrum minimizes the conduction band EMI and therefore eases the system design.

### ● Leading Edge Blanking

The PT2312E is designed to offer cycle-by-cycle current limiting in current mode control. The switch current is detected by a sense resistor  $R_{cs}$  at CS pin. Because of the parasitic parameters, there will be a voltage spike on  $R_{cs}$  at the moment of power MOSFET turn on, which will lead to the wrong control logic. An internal leading edge blanking (LEB) circuit chops off the sensed voltage spike and can be utilized to avoid this fault. Thanks to the LEB design, the RC filter on CS pin can be omitted.

### ● Adjustable Cable Compensation

A built-in cable drop compensation circuit is implemented to achieve good load regulation. An offset compensative current is added to FB pin, and the offset current changes with the load changes, thus the drop voltage on output cable

can be compensated.

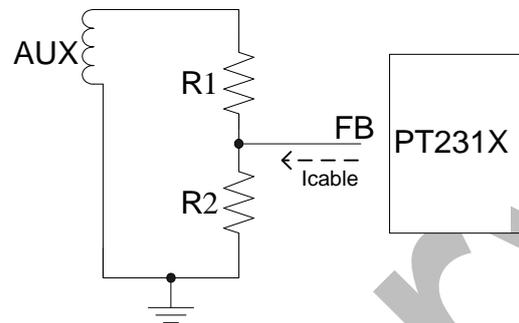


Figure 7. Cable compensation circuit

As shown in figure 7, set the cable compensation equation as

$$V_{ref} \left(1 + \frac{R1}{R2}\right) - I_{cable} \cdot R1 = \frac{N_{AUX}}{N_S} \cdot (V_O + \Delta V + V_{cable}) \quad (8)$$

Here,  $V_{ref}$  is the feedback voltage equal to 2.5V,  $\Delta V$  is the voltage drop on the output diode,  $V_{cable}$  is the voltage drop on the output cable and it is proportional to the load current. As the load current increases from no-load to full-load,  $I_{cable}$  will decrease from maximum (42.5 $\mu$ A) to 0, then the output voltage will increase. So a simple equation can be written as

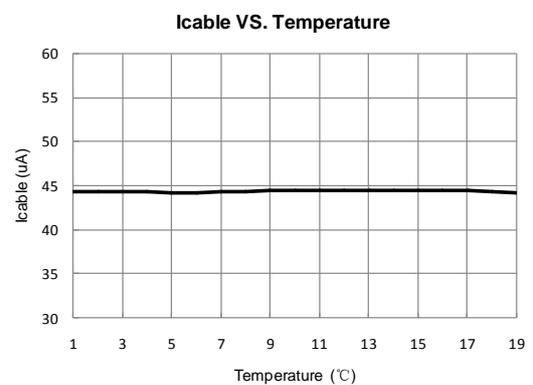
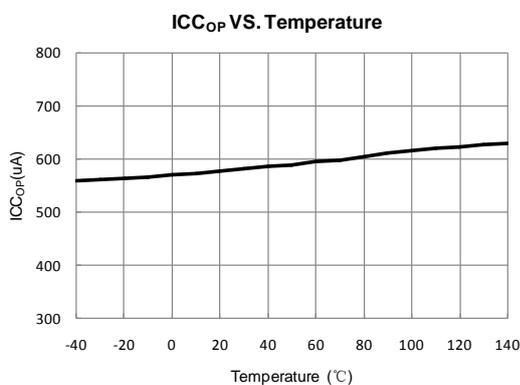
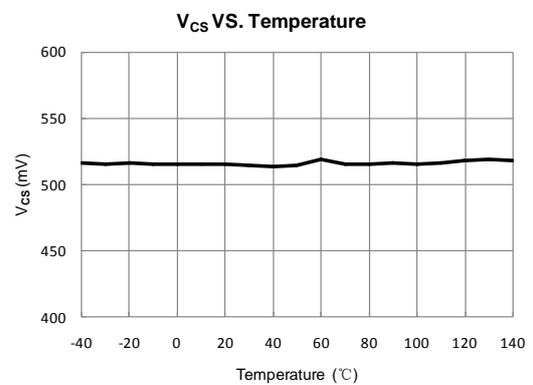
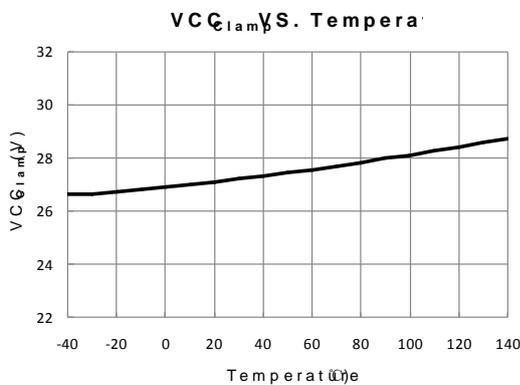
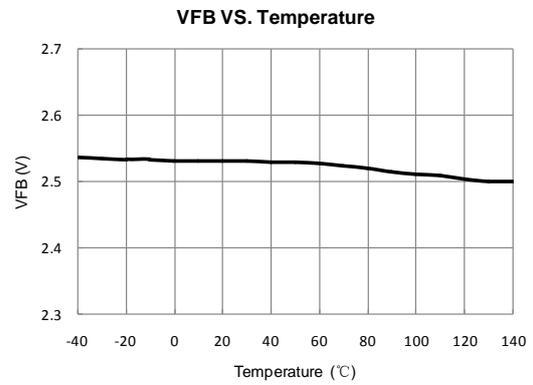
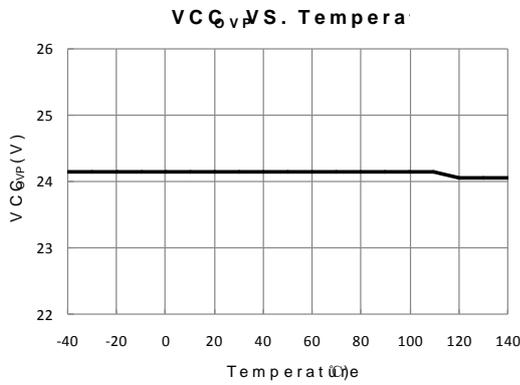
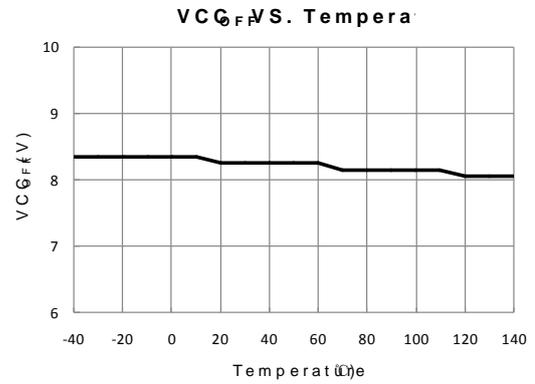
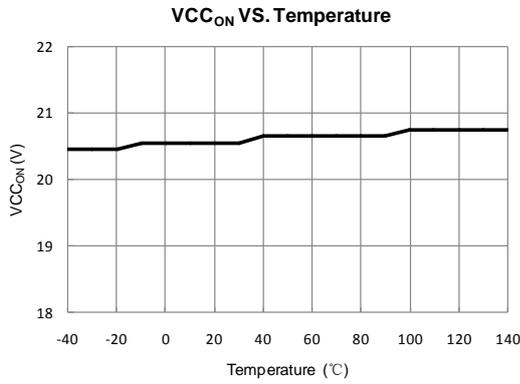
$$I_{cable} \cdot R1 = \frac{N_{AUX}}{N_S} \cdot V_{cable} \quad (9)$$

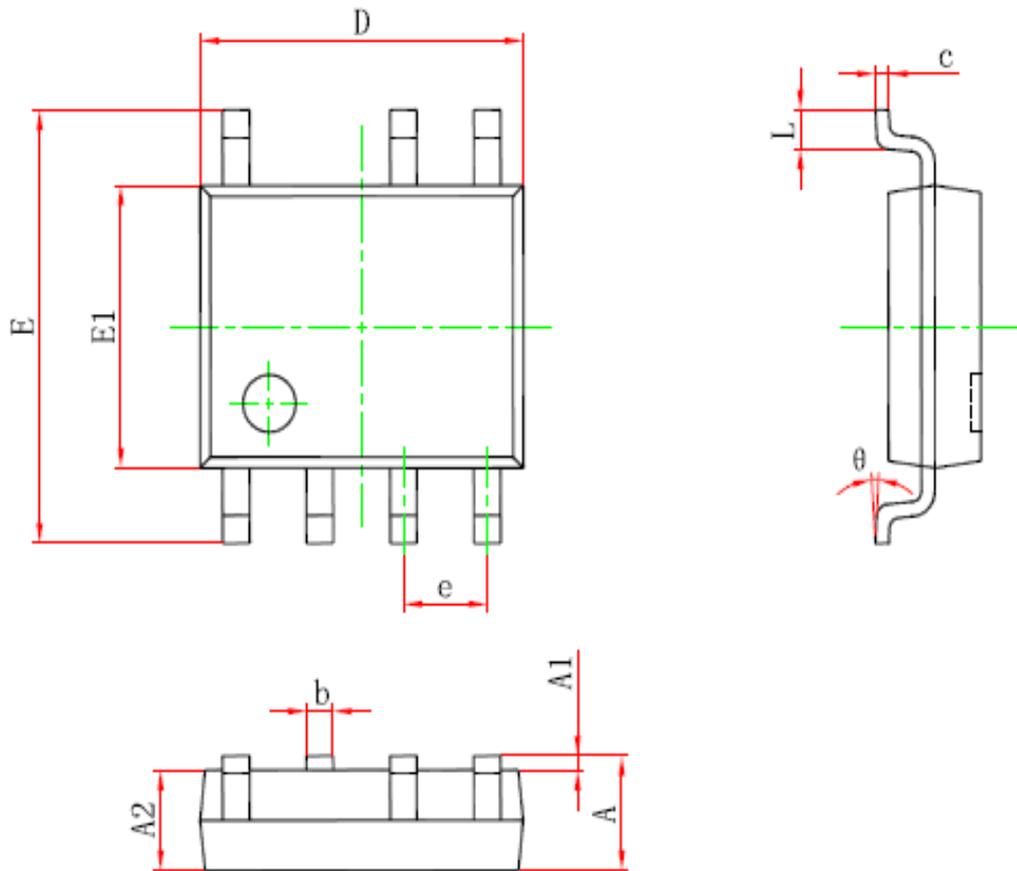
It is obvious that the compensative output voltage can be programmed by adjusting the resistance ( $R1$ ) of the divider on auxiliary winding.

### ● Protections

The PT2312E utilizes multi-protection to meet high reliability, such as cycle-by-cycle current limiting, output over voltage protection (OVP), VCC OVP/UVLO, CS resistor open/short protection, FB resistors open/short protection, over temperature protection etc.

## TYPICAL PERFORMANCE CHARACTERISTICS



**PACKAGE INFORMATION**
**SOP7 PACKAGE OUTLINE DIMENSIONS**


Symbol	Millimeters		Inches	
	Min	Max	Min	Max
A	1.350	1.750	0.053	0.069
A1	0.100	0.250	0.004	0.010
A2	1.250	1.650	0.049	0.065
b	0.330	0.510	0.013	0.020
c	0.170	0.250	0.007	0.010
D	4.700	5.100	0.185	0.201
E	3.800	4.000	0.150	0.157
E1	5.800	6.200	0.228	0.244
e	1.270(BSC)		0.050(BSC)	
L	0.400	1.270	0.016	0.050
θ	0°	8°	0°	8°

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